#### **■ FEATURES**

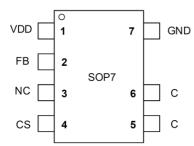
- ➤ Built-in High Voltage Power BJT
- Proprietary NC-Cap/super-QR/PSR (Quasi-Resonant & Primary Side Regulation) Control:
  - ◆ ±4% CC and CV Precision
  - ◆ Proprietary "Audio Noise Cancellation" Control
  - ◆ Built-in "Fast Dynamic Response" Control to Meet USB Charge Requirements
  - ◆ Proprietary "Zero-Output Startup" Control
  - ◆ Proprietary "Smart Output Short Protection"
  - without External Compensation/Filtering Capacitor Needed
  - ◆ Max. 50V Output for AC/DC LED Lighting

- Direct Drive of Low Cost BJT
- ➤ Proprietary Cable Drop Compensation
- ➤ Multi Mode Control
- > Frequency Jitter for Better EMI
- > Dynamic Base Drive for High Efficiency
- > Wide VDD Operating Range
- > 10.8V UVLO Hysteresis Window
- > Cycle-by-Cycle Current Limiting
- ➤ Leading Edge Blanking (LEB)
- > Built-in Soft Start, Pin Floating Protection
- > VDD UVLO, OVP, VOUT OVP, OLP, SCP

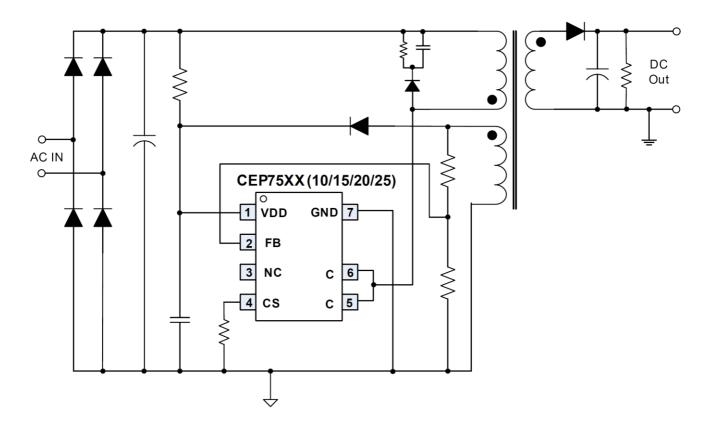
#### ■ APPLICATIONS

- > Battery chargers
- > Replaces linear transformer and RCC SMPS
- > Small power adapter
- > AC/DC LED lighting

#### ■ PACKAGE and SIMPLIFIED APPLICATION DIAGRAM



No.	Name	1/0	Description
1	VDD	Р	IC power supply pin.
2	FB	_	System feedback pin. This control input regulates both the output voltage in CV mode and output current in CC mode based on the fly-back voltage of the auxiliary winding.
3	NC		No connection.
4	CS	I	Current sense pin.
5-6	С	Р	High voltage power BJT collector pin.
7	GND	Р	Ground





#### ■ GENERAL DESCRIPTION

CEP75XX is a high precision, highly integrated DCM (Discontinuous Conduction Mode) Primary Side Regulation (PSR) power switch for offline small power converter applications. It has built-in power BJT, which can further lower system cost.

CEP75XX has built-in proprietary NC-Cap/super-QR/PSR control for CV control, which eliminates external compensation or filtering capacitor. It has built-in cable drop compensation function, which can provide excellent CV performance. The IC uses Multi Mode Control to improve efficiency and reliability and to decrease audio noise energy @ light loadings.

CEP75XX integrates proprietary "Audio Noise Cancellation" control for audio noise free operation. The IC has built-in "Fast Dynamic Response" control

to meet USB Charge requirements. CEP75XX also integrates proprietary "Zero-Output Startup" control to achieve startup when output is near zero voltage. The IC has proprietary "Smart Output Short Protection", which can protect the system with large leakage inductance when output is short circuit.

CEP75XX integrates functions and protections of FB Short Protection, Under Voltage Lockout (UVLO), VDD Over Voltage Protection (VDD OVP), VOUT Over Voltage Protection (VOUT OVP), Output Over Load Protection (OLP), Output Short Circuit Protection (SCP), Soft Start, Cycle-by-cycle Current Limiting (OCP), Pin Floating Protection, VDD Clamping.

CEP75XX is available SOP7 packages.

#### ■ Ordering Information

Part Number	Part Number Top Mark Package		Recommended Output Specification	Tape & Reel	
CEP7510SP7	C7510 YYWW XX	SOP7	5V1.0A	YES	
CEP7515SP7	C7515 YYWW XX	SOP7	5V1.5A	YES	
CEP7520SP7	C7520 YYWW XX	SOP7	5V2.0A	YES	
CEP7525SP7	C7525 YYWW XX	SOP7	5V2.4A	YES	

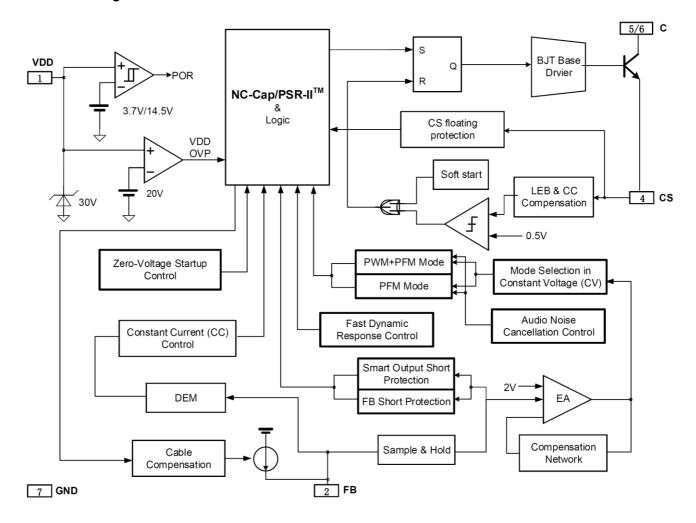
#### ■ Recommended Operation Conditions (Note 1)

Parameter	Value	Unit
Supply Voltage, VDD	7 to 16	V
Operating Ambient Temperature	-40 to 85	$^{\circ}$ C
Maximum Switching Frequency	70K	Hz

#### Absolute Maximum Ratings (Note 2)

Parameter	Value	Unit	
VDD DC Supply Voltage	22	V	
VDD DC Clamp Current		10	mA
CS voltage range		-0.3 to 7	V
FB voltage range		-0.7 to 7	V
Collector to Emitter Voltage	850	V	
Package Thermal Resistance (SOP-7)	$\theta$ JA	120	°C/W
Fackage Thermal Nesistance (SOF-1)	θјс	60	°C/W
Maximum Junction Temperature		150	$^{\circ}$
Operating Temperature Range	-40 to 85	$^{\circ}\mathbb{C}$	
Storage Temperature Range	-65 to 150	$^{\circ}\mathbb{C}$	
Lead Temperature (Soldering, 10sec.)	260	$^{\circ}\mathbb{C}$	
ESD Capability, HBM (Human Body Model)	3	kV	
ESD Capability, MM (Machine Model)		250	V

#### ■ Block Diagram



#### **■ ELECTRICAL CHARACTERISTICS**

(T<sub>A</sub> = 25°C, VDD=14V, if not otherwise noted)

Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit			
Supply Voltage (VDD) Section									
I_Startup	VDD Start up Current	VDD =UVLO(OFF)-1V		1	3	uA			
I_VDD_Op	Operation Current	VDD=14V		0.8	1.5	mΑ			
UVLO(OFF)	VDD Under Voltage Lockout Exit (Startup)		13.5	14.5	15.5	V			
UVLO(ON)	VDD Under Voltage Lockout Enter		3.4	3.7	4	V			
VDD_OVP	VDD Over Voltage Protection trigger		18	20	22	V			
T_Softstart	Soft Start Time			2		mSec			
Feedback In	put Section(FB Pin)								
V <sub>FB</sub> _EA_Ref	Internal Error Amplifier(EA) reference input		1.98	2.0	2.02	V			
V <sub>FB</sub> _DEM	Demagnetization comparator threshold			25		mV			
Tmin_OFF	Minimum OFF time			2		uSec			
T <sub>max</sub> _OFF	Maximum OFF time			3.3		mSec			
V <sub>FB</sub> _Short	Output Short Circuit Threshold			1.2		V			



T <sub>FB</sub> _Short		Tim				65		mSec		
V <sub>FB</sub> _OVP		Thre	put Over Voltage Protection eshold			2.8		V		
Тсс/Тоем			o between switching period in mode and demagnetization			2				
Icable_ma	ax	Max	Cable compensation current			60		uA		
Curren	t Sen	se Ir	put Section (CS Pin)		•		•	-		
T_blank	ing	CS	Leading Edge Blanking Time			500		nSec		
Vth_OC	_max	Max	c. Current limiting threshold		490	500	510	mV		
T <sub>D</sub> _OC		Ove Dela	er Current Detection and Control ay			100		nSec		
Power	BJT S	Secti	on							
	C75	10		Ic=1mA	450					
BVceo	C75	15	Collector-emitter breakdown	Ic=10mA	480			V		
D V CEO	C75	20	voltage	Ic=10mA	450					
	C75	25		Ic=10mA	450					
	C75	10		Ic=1mA	750			V		
ВУсво	C75	15	Collector-base breakdown	Ic=1mA	850					
<b>D</b> V CBO	C75	20	voltage	Ic=1mA	800					
	C7525			Ic=1mA	800			1		
	C75	10		I <sub>E</sub> =1mA	11					
BV <sub>EBO</sub>	C75	15	Emitter-base breakdown	I <sub>E</sub> =1mA	11			V		
<b>D</b> A EBO	C75	20	voltage	I <sub>E</sub> =1mA	9					
	C75	25		I <sub>E</sub> =1mA	9					
	C75	10		Ic=0.1A, I <sub>B</sub> =20mA			0.6			
	C75	15		Ic=0.1A, I <sub>B</sub> =20mA			0.5			
V <sub>CE(sat)</sub>	073	10	Collector-emitter saturation	Ic=0.5A, Iв=0.1A			8.0	V		
V CE(sat)	C75	20	voltage	Ic=1A, Iв=0.2A			0.4	V		
	C7525			Ic=1.0A, I <sub>B</sub> =0.2A			0.4			
	013	20		Ic=2.0A, I <sub>B</sub> =0.5A			0.6			
	C75	10		Vce=20V, Ic=20mA	20		30			
Hfe	C75	15	DC current gain	Vce=5V, Ic=0.1A	20		30	]		
піе	C75	20	Do current gain	Vce=6.5V, Ic=0.2A	20		30			
	C75	25		Vce=5V, Ic=1A	15		25			

**Note 1.** Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**Note 2.** The device is not guaranteed to function outside its operating conditions.



### NC-Cap/super-QR/PSR CV/CC Power Switch

#### **OPERATION DESCRIPTION**

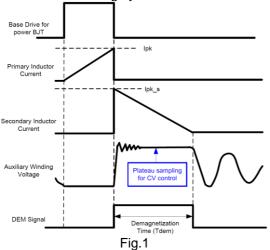
CEP75XX is a high performance, highly integrated DCM (Discontinuous Conduction Mode) Primary Side Regulation (PSR) power switch. The built-in high precision CV/CC control makes it very suitable for offline small power converter applications.

#### > PSR Technology Introduction

Assuming the system works in DCM mode, the power transfer function is given by

$$P = \frac{\eta}{2} \times L_m \times I_{pk}^2 \times f_S = V_o \times I_o \qquad \text{(Eq.1)}$$
 In the equation above, P is output power, Vo and

lo are system output voltage and current respectively,  $\eta$  is system power transfer efficiency, Lm is transformer primary inductance, fs is system switching frequency, lpk is primary peak current in a switching cycle. The following figure illustrates the waveform in a switching cycle.



In the figure shown above, the IC generates a demagnetization signal (DEM) in each switching auxiliary winding. through demagnetization time for CV/CC control. In DCM mode, Tdem can be expressed as;  $\frac{V_o}{L_m} \times T_{dem} = \frac{N_S}{N_P} \times I_{pk}$ 

$$\frac{V_o}{L_m} \times T_{dem} = \frac{N_S}{N_P} \times I_{pk}$$
 (Eq.2)

In Eq.2, Np and Ns are primary and secondary winding turns respectively.

Combined with Eq.1 and Eq. 2, the average output current can be expressed as:

$$I_o = \frac{\eta}{2} \times I_{pk} \times \frac{N_P}{N_S} \times f_S \times T_{dem} \qquad \text{(Eq.3)}$$
  $\spadesuit$  CC (Constant Current) Control Scheme

From Eq.3, it can be easily seen that there are two ways to implement CC control: one is PFM (Pulse Frequency Modulation), the control scheme is to keep lpk to be constant, let the product of Ts and Tdem (fs\*Tdem) to be a constant. In this way, lo will be a value independent to the variation of Vo,Lm, and line input voltage. Another realization method is PWM duty control, the control scheme is to keep fs to be constant, let the product of Tdem and lpk (Tdem\*lpk) to be a constant, in another words, by modulating system duty cycle to realize a constant lo independent to the variation of Vo, Lm and line voltages.

CEP75XX adopts PFM for CC control, the product of Ts and Tdem is given by

$$f_S \times T_{dem} = 0.5 \tag{Eq.4}$$

#### ◆ CV (Constant Voltage) Control Scheme

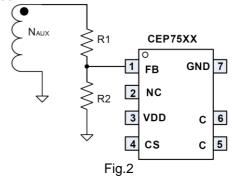
CV control should sample the plateau of auxiliary winding voltage in flyback phase, as shown in Fig.1 The CV control has many implementations, for example, PWM, or PFM, or a combination of both one. In CEP75XX, the CV control adopts proprietary multi mode control, as mention below.

The output voltage can be sensed via the auxiliary winding. During MOSFET turn-off time, the energy stored in the primary winding is transferred to the secondary side. The auxiliary voltage reflects the output voltage as shown in Fig.1 and it is given by

$$V_{AUX} = \frac{N_{AUX}}{N_S} \times (V_{\theta} + \Delta V)$$
 (Eq.5)  
V indicates the voltage drop of t

Where  $\triangle$  V indicates the voltage drop of the output diode.

As shown in Fig.2, via a resistor divider connected between the auxiliary winding and FB (pin 1), the auxiliary voltage is sampled at the middle of the demagnetization and it is hold until the next sampling. The sampled voltage is compared with reference voltage (typical 2.0V) and the difference is amplified. The error amplifier output reflects the load condition and controls the switching off time to regulate the output voltage, thus constant output voltage can be achieved.



#### Startup Current / Startup Control / Operating Current

Startup current of CEP75XX is designed to be very low (typically 1uA) so that VDD could be charged up above UVLO(OFF) threshold level and device starts up quickly. The operating current in CEP75XX is as small as 0.8mA (typical). The small operating current results in higher efficiency and reduces the VDD hold-up capacitance requirement.



### NC-Cap/super-QR/PSR CV/CC Power Switch

#### > NC-Cap/PSR Introduction

#### ◆ ±4% Precision CV/CC Performance

CEP75XX can achieve less than  $\pm$  4% variation of CC/CV precision due to the built-in CV accuracy improvement and CC line and load compensation, as shown in Fig.3.

#### ◆ Proprietary "Audio Noise Cancellation" Control

CEP75XX has a proprietary "Audio Noise Cancellation" control, which can achieve audio noise free operation in the whole loading range.

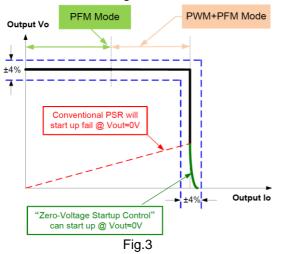
#### ◆ Built-in Fast Dynamic Response Control to Meet USB Charge Requirements

In CEP75XX, a fast dynamic response control is integrated to improve system dynamic response performance, thus the charger system can meet the USB charge requirements.

#### ◆ Proprietary Zero-Output Startup Control

Conventional PSR system may suffer startup failure when output voltage is near zero voltage, which means that there is a gap between OCP (CC point in PSR CV/CC system) and full loading. Larger OCP gap causes larger system cost.

In CEP75XX, a proprietary "Zero-Output Startup Control" is adopted to achieve successful startup @ Vout≈0V, as shown in Fig.3.



### **♦** Smart Output Short Protection

The output short circuit protection of conventional PSR system is based on the coupling between auxiliary winding and secondary winding. When output is short, the auxiliary winding cannot provide enough energy to the IC any more. In this way, the system will enter into auto-recovery mode protection.

However, the IC may be wrongly supplied if the leakage inductance of the primary winding is large enough.

In CEP75XX, if output short circuit occurs, the IC will detect the situation and enter into auto-recovery mode protection.

### ♦ No External Compensation/Filtering Capacitor Needed

CEP75XX uses a proprietary control to eliminate external compensation capacitor, which can simplify system design and lower system cost.

#### **◆ Maximum 50V Output for LED Lighting**

CEP75XX can support maximum 50V output, which can be used in AC/DC LED lighting.

#### Multi Mode PSR Control for High Reliability , High Efficiency

Conventional pure PFM controlled PSR system may suffer transformer saturation issue when heavy loading. In CEP75XX, a proprietary multi mode control is adopted to suppress this issue, as shown in Fig.3. Around the full load, the system operates in PWM+PFM mode, which improve the system reliability. Under normal to light load conditions, the IC operates in PFM mode to achieve excellent regulation and high efficiency.

#### > Soft Start

CEP75XX features an internal 2ms (typical) soft start that slowly increases the threshold of cycle-by-cycle current limiting comparator during startup sequence. Every startup process is followed by a soft start activation.

#### ➤ Leading Edge Blanking (LEB)

Each time the power BJT is switched on, a turn-on spike occurs across the sensing resistor. To avoid premature termination of the switching pulse, an internal leading edge blanking circuit is built in. During this blanking period (500ns, typical), the cycle-by-cycle current limiting comparator is disabled and cannot switch off the base driver.

#### > Minimum and Maximum OFF Time

In CEP75XX, a minimum OFF time (typically 2us) is implemented to suppress ringing when BASE drive is pull off. The maximum OFF time in CEP75XX is typically 3.3ms, which provides a large range for frequency reduction. In this way, low standby power can be achieved.



### NC-Cap/super-QR/PSR CV/CC Power Switch

#### > Proprietary Cable Drop Compensation

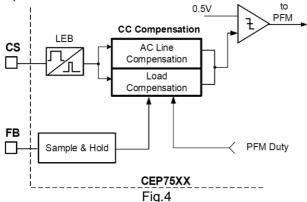
CEP75XX has a proprietary built-in cable voltage drop compensation block which can provide a constant output voltage at the end of the cable over the entire load range in CV mode.

#### > Pin Floating Protection

In CEP75XX, if pin floating situation occurs, the IC is designed to have no damage to system.

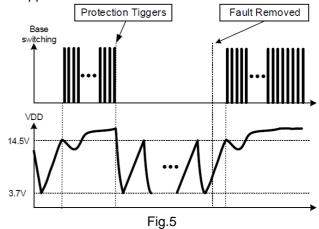
#### > Built-in Load and AC Line CC Compensation

In conventional PSR system, the output CC (Constant Current) point can vary with output and AC line voltage. In CEP75XX the IC has built-in blocks to compensate the variation, as shown in Fig4. The IC can adjust CC point based on sensed output voltage and PFM duty. In this way, CC accuracy can be improved.



#### > Auto Recovery Mode Protection

As shown in Fig.5, once a fault condition is detected, switching will stop. This will cause VDD to fall because no power is delivered form the auxiliary winding. When VDD falls to UVLO(on) (typical 3.7V), the protection is reset and the operating current reduces to the startup current, which causes VDD to rise. However, if the fault still exists, the system will experience the above mentioned process. If the fault has gone, the system resumes normal operation. In this manner, the auto restart can alternatively enable and disable the switching until the fault condition is disappeared.



### > VDD Over Voltage Protection

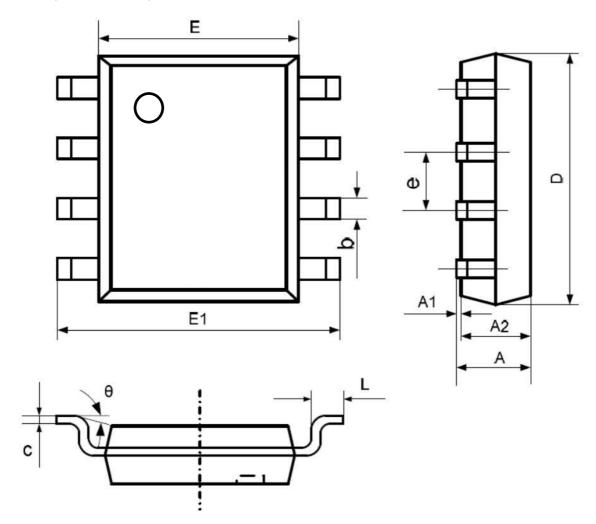
VDD OVP is implemented in CEP75XX and it is a protection of auto-recovery mode. If VDD is higher than 20V (TYP), VDD OVP protection will be trigged and the switch is turned off until VDD is lower than UVLO(on) (typical 3.7V).

#### > VOUT Over Voltage Protection

If FB voltage during the sampling time is higher than OVP voltage (typical 2.8V), the counter will start to work. VOUT OVP protection will be trigged when the counter is counted to 3 continuously, and the switch is turned off until VDD is lower than UVLO(on) (typical 3.7V).

### **■ PACKAGE OUTLINE**

SOP-8(SOP-7 covered) PACKAGE OUTLINE AND DIMENSIONS

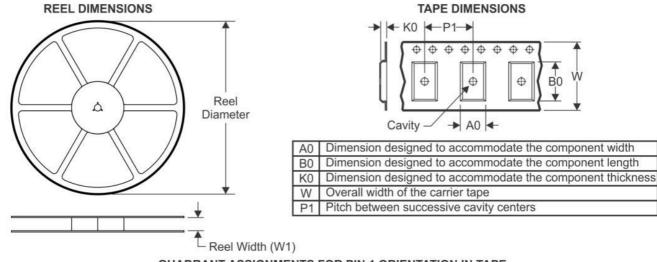


SYMBOL		ISION IN METERS	DIMENSION IN INCHES			
	MIN	MAX	MIN	MAX		
А	1.350	1.700	0.053	0.067		
A1	0.000	0.100	0.000	0.004		
A2	1.350	1.550	0.053	0.061		
b	0.330	0.510	0.013	0.020		
С	0.170	0.250	0.007	0.010		
D	4.700	5.100	0.185	0.200		
Е	3.800	4.000	0.150	0.157		
E1	5.800	6.200	0.228	0.244		
е	1.270	TYP	0.050 TYP			
L	0.400	1.270	0.016	0.050		
θ	0°	8°	0°	8°		

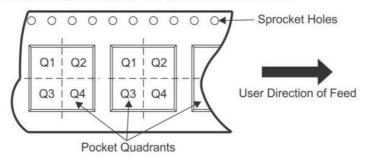


### NC-Cap/super-QR/PSR CV/CC Power Switch

#### TAPE AND REEL INFORMATION



#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

Device		Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)
CEP7510SP7		C7510	7 40	4000	330.0	12.8	6.4	5.2	2.5	8.0	12.0
CEP7515SP7	000.7	C7515									
CEP7520SP7	SOP-7	C7520		4000							
CEP7525SP7		C7525									